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Magnetic binary encoding system based on 3D printing and GMI detection prototype

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ABSTRACT

In this work, the feasibility of a magnetic binary encoding system using 3D printing technology is analyzed. The study has a double interest, that is, the possibility of printing a 3D piece that contains the codified information and the development of a system for its decoding. For this purpose, magnetic nanoparticles (magnetite Fe_3O_4) were embedded in a polymeric matrix of Polylactic Acid (PLA) and Poly- ϵ -caprolactone (PCL). Similar to a conventional barcode, a rectangular piece with an alternating pattern of strips with absence (only polymer) and a 5 wt% of embedded magnetic nanoparticles was 3D printed employing the Fused Deposition Modelling technique (FDM). The information was decoded by means of a Giant Magnetoimpedance (GMI) sensor-based prototype, by scanning the surface of the piece and measuring the changes in the magnetic field. As sensor nucleus, an amorphous soft magnetic wire of nominal composition ($Co_{0.94}$ Fe_{0.06})_{72.5} Si_{12.5} B₁₅ was employed. The decoding prototype incorporates a homemade electronic sensor interface that permits, at the time, the GMI sensor excitation and the subsequent signal conditioning to optimize its response. The output signal enables the detection of the magnetic nanoparticles and the magnetic decoding of the encoded information ("1" and "0", presence or absence of the magnetic nanoparticles, respectively).

1. Introduction

The research of novel and functional magnetic materials has recently become a relevant research topic [1]. As a result of their relevant features related to their, generally speaking, fast, long-range and precise response in different environments [2], these materials have been implemented in new devices and applied in emerging technologies such as controlled motion [3], [4], mechanical reinforcement [5], [6], strain sensors [7], [8], and biomedical applications [9], [10] among others. In this sense, different magnetic devices have been developed based on the combination of distinct materials and the use of sophisticated structural geometries, showing an optimal performance under different conditions at various applications [2]. Some examples of this kind of devices can be found in the bibliography, for example, transformers [11], microfluidic platforms [12] and magnetic actuators [8], [13], etc. Anyway, the development of new structures requires novel manufacturing

procedures that allow overcoming the constraints that exhibit the traditional fabrication strategies, mainly based on the predominant subtractive manufacturing processes. Under these principles, 3D printing represents a promising technique for a rapid and self-customized manufacturing of these complex magnetic (and non-magnetic) structures or geometries [2], [14]. This interdisciplinary technology consists of a two-step process. Initially, a computer-assisted 3D model is developed. Then the final printing process is carried out leading to a final 3D object. This stage consists of the successive deposition of thin layers of materials upon each other [15], [16]. Different materials are used in 3D printing, namely plastic, resins, rubbers, biomaterials, ceramics, glass, metals, etc [17]. So, among others, the advantages of this technique include the development of devices with complex internal structures (hollow and filled parts), the use of a wide variety of materials (single or composite), rapid and low-cost prototyping and a high level of personalization, allowing a fast adaptation to the requirements of the desired

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In this work, the feasibility of 3D printed magnetic composites (Fused Deposition Modelling, FDM) for binary encoding has been analyzed. To this end, magnetic composites that consist of nanometric commercial Fe₃O₄ particles (MNPs), i.e., 50–100 nm, were embedded in a semi-crystalline, biocompatible and biodegradable polymeric blend made of a Polylactic Acid (PLA) and Poly- ε -caprolactone (PCL). A rectangular-shaped piece was printed (similar to a visual conventional barcode), being formed by 4 strips with the same dimensions. An alternating pattern was chosen, where a polymeric strip (no MNPs) was followed by another with embedded MNPs and so on. Consequently, the encoding principle is derived from the capacity of detecting the MNPs presence, "1", or their absence, "0" when a magnetic sensor scans over the printed piece surface.

Derived from the low MNPs concentrations involved, a detection system based on the Giant Magnetoimpedance effect (GMI) was proposed. This effect consists of the huge changes in the high-frequency electrical impedance, Z, experienced by a soft magnetic conductor under an external magnetic field [18], such as the one generated by the MNPs. Its employment is justified by the larger sensitivity exhibited when compared to other magnetic detecting principles [19]. In those terms, previous GMI-based sensors for MNPs detection have been successfully developed [20], [21], [22]. More concretely, we have previously analyzed this topic using different geometries as ribbons [23], [24] and wires for the contactless and reusable detection [25] of the magnetic particles. In this work, a detection system is analyzed based on the voltage (amplitude), V, variations $(V \propto Z)$ underwent by an amorphous wire of nominal composition (Co_{0.94}Fe_{0.06})_{72.5}Si_{12.5}B₁₅, while scanning over the 3D barcoded piece. The final prototype is completed by the addition of a homemade electronic sensor interface that permits, simultaneously, the sensor excitation and the signal conditioning to optimize its response.

2. Materials and methods

2.1. Preparation of magnetic composites

The fabrication of 3D printable filaments was initially addressed. Thus, a concentration of 5 wt% of iron oxide magnetic nanoparticles (Fe₃O₄) were embedded in a polymeric matrix with 90% PCL and 10% PLA. Commercial Fe₃O₄ MNPs (Sigma Aldrich, 637106), i.e., 50–100 nm



Fig. 1. Hysteresis loop at room temperature of commercial magnetite MNPs. Inset. Enlargement of the low magnetic field region for the different hysteresis loops measured at different maximum applied fields: 1.2 MA/m, 100 kA/m and 60 kA/m.

were used for the composite fabrication. Fig. 1 shows the hysteresis loop measured with a homemade Vibrating-sample magnetometer (VSM) under a maximum applied field of 1.2 MA/m. As can be observed in the inset of Fig. 1, a non-zero value of the remanent magnetization (\sim 5.5 emu/g) was found for this situation. Although the MNPs are within the limit of superparamagnetic behavior [26], the occurrence of particle size dispersion caused that some nanoparticles could display a certain magnetization even at zero applied external magnetic field.

The solution casting method was used for the synthesis of homogeneous magnetic composites [27]. The following reagents were used: Polylactic acid (PLA, average molecular weight [Mw]≈ 144.000, Natureworks, USA), Polycaprolactone (PCL, average molecular weight $[Mw] \approx 50.000$, Polymorph, UK), magnetic iron oxide nanoparticles and dichloromethane (DCM) solvent both purchased from Sigma Aldrich. Briefly, PLA was dissolved in 300 mL of DCM with mechanical stirring at 30 °C. As the PLA was dissolved, PCL was added. To promote a homogeneous mixture between both polymers, the solution was stirred vigorously for 2 h. Then magnetite nanoparticles were added. Stirring was kept for one more hour. Finally, the temperature was increased to 45 °C to facilitate the evaporation of DCM solvent, obtaining a composite precursor with a foam texture (see Fig. 2a). The final step was the PLA/PCL/MNPs filament extrusion. The fabrication of magnetic filaments was carried out using an extruder (FelFil Evo) with a 1.75 mm circular cross-section nozzle. The extrusion temperature was 80 °C. An example of the printable 3D filament is shown in Fig. 2b.

2.2. Printing of 3D piece

With the developed printable filament, a magnetic barcode piece was 3D printed by FDM technique. The piece was fabricated by NAITEC, Technological Centre of Mobility and Mechatronics of Navarre, using a Raise 3D E2 printer with two print heads (dual extruders), one for the magnetic filament and the other for the pure polymeric matrix. As shown (see Fig. 3), the rectangular-shaped piece was comprised of 4 strips with an alternating pattern of absence and presence of MNPs. All strips exhibited the same dimensions, namely, 10 mm of length (*x*-direction), 4 mm of width (*y*-direction) and 2 mm of thickness (*z*-direction).

2.3. GMI Sensor treatment

The GMI sensing element is based on a soft magnetic amorphous wire (Co_{0.94} Fe_{0.06})_{72.5}Si_{12.5}B₁₅ obtained by "in-rotating-water-quenching" [28]. Initially, a 5 cm in length and 140 μm of diameter wire was sequentially current annealed (t = 5 min, $i = 19.5 \text{ A/mm}^2$) and submitted to torsional strain, ranging from $\xi = 0$ to 10π rad/m, to induce a helical anisotropy on the sample [29]. Under a voltage divider configuration, the evolution of the sensor voltage, V, was characterized as a function of the external DC magnetic field, H, after each thermal-torsional treatment. The magnetic field was applied along the wire axis and generated by a pair of Helmholtz coils $(H = (583 \frac{Am^{-1}}{A} \bullet I(A); I \text{ electrical current coil of radius } R = 4.5 \text{ cm}).$ The sensor was excited under conditions that optimized the GMI effect, namely, under a frequency of f = 100 kHz and peak-to-peak current intensity, I_{pp} = 20 mA. To this end, a standard signal generator (Standford Research Systems DS 345) was employed, being the sinusoidal AC signal amplitude measured with an AC current probe (Tektronix P6021). Voltage variations, V, were registered with a commercial lock-in amplifier (Standford Research Systems SR 844) as a function of H. It must be pointed out that the treatment of the wire was performed to enhance its GMI sensitivity in the low magnetic field region. As Fig. 4 demonstrates, due to the reinforcement of the helical anisotropy with the torsion angle during the current annealing, the position (applied Hfield) of the maximum of the GMI voltage shifts towards higher magnetic field values, \pm 90 A/m. Moreover, a progressive enhancement of the



Fig. 2. a) Foam 3D filament precursor after chemical synthesis and b) 3D printable filament with MNPs embedded (5% wt).



Fig. 3. Barcode patterned 3D printed piece.

sensitivity (higher slope) and an enlargement of the *H* range with linear sensor response, resulted in increasing applied torsional strains (see inset of Fig. 4) leading to a final mean sensitivity of 2×10^{-3} V/Am⁻¹ (or 0.16 V/Oe). For sensing purposes, a piece of 10 mm twisted $\xi = 10\pi$ rad/m of the annealed wire was cut to perform the MNPs detection stage through the 3D piece surface scan.

2.4. Electronic sensor interface

The detection system consisted of an analog interface for both the excitation of the sensor and the signal conditioning of the GMI voltage. The complete scheme is shown in Fig. 5. The homemade designed electronic system was implemented in a breadboard using a single supply voltage of 5 V and a virtual ground *Vref* at 2.5 V. Operational amplifiers AD823 of Analog Devices were employed. The system consisted of three different blocks: the excitation signal generation, the sensing stage and an envelope detector that provided the final output *DC* voltage, *V*_{out}. A LM555 timer was used for generating a 100 kHz square signal, which was subsequently integrated to obtain a triangular waveform. This triangular signal is converted to current (*f* = 100 kHz and *I*_{pp})



Fig. 4. Evolution of the 5 cm wire voltage, *V*, with external magnetic field, *H* in the initial as-quenched state and after successive current annealing treatment under different torsional strain, ξ .



Fig. 5. Complete electronic sensor interface, including excitation signal generation and envelope detection circuit to obtain the output DC voltage.

= 20 mA) and applied to the sensor by using a transimpedance amplifier configuration. The excitation current was set by a variable resistor that controls the voltage gain of the opamp-based integrator.

The sensor output voltage, V_{-} , $(V_{-} \propto (I_{pp} \bullet Z))$ was amplified utilizing an *AC*-coupled non-inverter amplifier and the amplitude of the resulting voltage is measured employing a simple envelope detector, using 1N4148 diodes. The non-inverting amplifier has a double function: on the one hand, removing the possible *DC* offsets previously introduced by the setup, and on the other hand, ensuring an output signal amplitude large enough for dealing with the dropout voltage of the passive diodes.

Note that a replica of the main branch was added in parallel to the sensing circuit. This extra circuitry aimed to obtain a reference voltage, V_{-} to be subtracted from the output voltage of the sensor branch. In this way, the common-mode undesired effects introduced by the setup were rejected at the output. Moreover, the gain of the transimpedance amplifier of the reference branch can be adjusted to partially compensate the output measured amplitude, given that the amplitude variations introduced by the GMI sensor used to be small compared to the total voltage swing. Thus, higher voltage variations can be obtained at V_{out} by introducing an extra gain in the final subtractor.

The response of the final decoder system (V_{out}) was initially calibrated under controlled conditions. For that, the 10 mm annealed wire was located at the center of the Helmholtz coils and submitted to H field. For comparison, the GMI voltage of the 10 mm annealed wire was similarly characterized under the previous voltage divider (V) configuration. The V analysis (inset of Fig. 6) shows a displacement of the maxima towards higher magnetic fields, H (-378 A/m and 368 A/m)respecting the initial 5 cm in length wire. Besides, a clear diminution in the mean sensitivity of the sensor was produced, concretely almost two orders of magnitude, reducing its value to $6\times 10^{-5}~V/Am^{-1}$ (or 4.8 $\times 10^{-3}$ V/Oe). These effects can be ascribed to the well-known existence of a critical length in the GMI response. This fact is the result of the increasing role of the closure domains that appear at the ends of the sample after cutting the wire, causing a modification in the magnetic domain structure of the sample [30], [31] and consequently, in its characteristic transverse magnetic permeability. This noticeable decrease represents a trade-off between sensitivity and suitable dimensions for potential applications. In this sense, longer wires exhibit a much larger capacity of detection, however, the shrinkage of components is a required requisite that cannot be disregarded for practical purposes. In any case, further analyses for sensing wire optimization should be addressed in the search for sensitivity improvement i.e. differential sensing configuration [32], non-linear GMI effects, use of GMI



Fig. 6. Electronic sensor interface output signal, V_{out} , versus applied magnetic field, *H* for the 10 mm annealed wire. Inset: Wire response, *V*, versus *H* under voltage divider configuration.

phase-based detectors, since they exhibit larger sensitivities when compared to GMI magnitude-based detectors, even at low samples lengths, i.e. 1 cm [33–36], etc. As expected, the previously mentioned displacement in the voltage peaks towards higher *H* was also found for the calibrating curve (see Fig. 6). This curve, in the low field region, displays a continuous increase of *V*_{out} reching a maximum value around $H \approx 375$ A/m. Under the settled electronics gain parameters, a final mean sensitivity of 3×10^{-3} V/Am⁻¹ (or 0.24 V/Oe) for the selected operation point was found in the final decoder system.

2.5. Scanning of the 3D printed piece: decoding stage

Then, the sensing wire was fixed to a standard glass cover used in microscopy with a square-shaped form, 15 mm side and 0.2 mm thickness. The ensemble was located over the 3D printed piece with the sensor axis along the x-direction (see Fig. 3). The scanning was performed along the y-direction with a step of 0.5 mm from the initial position at y = 0mm until the final position, y = 18mm, beyond the piece upper limit at y = 16mm (see Fig. 3). The movement was transmitted through an adapted commercial 3D printer motor (Artillery Sidewinder x1). A computer using LabView 2014 was employed both for the barcode scanning and the acquisition of the sensor output signal, V_{out} , through a multimeter HP34401A (see Fig. 7). Two different situations were analyzed: i) detection of the MNPs for the as-printed piece and ii) after the application of an external magnetic field, H_{elec} , applied to the whole piece along the x- direction. For the initial configuration i), different distances z were used. The lowest analyzed distance was z = 0.3mmcorresponding to the glass cover thickness plus an extra distance to avoid physical contact. This is the employed distance when no further detail is given. For ii) the magnetic field, $H_{elec} = 0 \frac{kA}{m}$, $60 \frac{kA}{m}$ and $100 \frac{kA}{m}$, was generated by an electromagnet (Applied Magnetics Laboratory, 4H2-45). Notice that this pre-magnetizing field was applied before the scanning measurements.

In order to analyze the possible effect of the direction of the MNPs field on the decoder system, an external and positive magnetic field close to 40 A/m was applied during the scanning stage. From that point (labeled as A in Fig. 6), the application of a positive or negative field would produce, respectively an increase or a decrease in V_{out} (see arrows in Fig. 6). Attending to Fig. 1, this magnetic field does not affect the MNPs since much higher fields are required for the change in their magnetization.

3. Results and discussion

First, the capacity of detection of the proposed decoder system



Fig. 7. Scheme of the complete experimental setup.

(sensor nucleus and electronic sensor interface) was checked. In this analysis, the distance z was kept constant at 0.3 mm. This starting configuration represents what hereinafter is called the "initial position", 0°. Moreover, for the sake of clarity, all shown figures incorporate two grey bars at the position of the strips filled with MNPs. can be observed in Fig. 8a, starting from $y \approx 0mm$, a clear increase of V_{out} happened as the sensor approached to the MNPs area, in such a way that, a maximum is obtained in the center of the first MNPs strip (labeled as 1 in Fig. 8a). Then, a gradual decrease in V_{out} resulted until reaching a relative minimum in the region between the magnetic strips ($y \approx 10mm$, center of the polymeric strip). An equivalent behavior was found beyond that position for the second MNPs strip (labeled as 2 in Fig. 8a. It is important to note that Vout displays larger values at the center of the empty polymeric strip between magnetic bars, (y \approx 10mm, V_{out} = 2.0987V) than at the center of the first polymeric strip, ($y \approx 2mm$, $V_{out} = 2.0963V$), and the final measured point, ($y \approx 18mm$, $V_{out} = 2.0947V$, although all related positions are equally separated from the closest magnetic strip. This situation can be interpreted in terms of the capacity of the GMI sensor to sense the magnetic field of the embedded MNPs at a certain distance. While relative voltage differences of $\Delta V = 5.5mV$ and 4mV(see Fig. 8a) were respectively found between the two maxima and the minimum at $y \approx 10mm$, higher variations close to 8 mV were observed between those maxima and the adjacent minima situated at $y \approx 2mm$ and at $y \approx 18mm$. Attending to the device sensitivity, these values corresponded to relative variations of the sensed magnetic field of $\Delta H =$ 1.8 A/m, 1.3 A/m and 2.7 respectively, confirming that the sensor experienced a larger magnetic difference when the sensor is only under the "effect" of one magnetic strip.



Fig. 8. Variation of the output signal, V_{out} , versus position, y, of the sensor: (a) at the initial position, 0° , and (b) after being rotated 180° the coded strips.

Then, the sample was rotated 180° (see Fig. 8b). Accordingly, the magnetic field direction and the position of the MNPs strips were altered respecting the initial configuration (see arrows in Fig. 8). Concretely, strip 2 was placed at the position y = 0 mm and initially detected in the scan. As it is shown, two minima in Vout were obtained at the center of the MNPs strips ($y \approx 2mm$ and $y \approx 10mm$ respectively), while a relative maximum was found in the region between bars ($\gamma \approx 6mm$). Similarly to the previous case, relative voltage variations of $\Delta V = 6.3 mV$ and 4.8mV ($\Delta H = 2.13$ A/m and 1.6 A/m) were found between the minima at and the maximum between magnetic bars voltages values. Moreover, it must be noted the rapid increase in V_{out} ($y \approx 16mm$), linked to the fact that the sensing wire only experienced the effect of one magnetic strip. This fact firstly reveals the feasibility of encodingdecoding information procedure based on 3D printed pieces technology, and secondly, that the prototype is sensitive to the direction of the magnetic field of the particles. Thus, magnetic polarity can also be successfully used to encode information in 3D printed objects. It is important to mention that during the decoding process there is no need of applying an external magnetic field to magnetize the MNPs in the piece, easing the development of a potential final prototype. However, when polarity is analyzed, a biasing field (40 A/m) is required. It is important to note that its intensity was quite low to avoid any change in the magnetization of the MNPs, but enabled the setting of the sensing in an optimal operation point to detect the MNPs field polarity.

One of the main advantages of magnetic codification is the possibility of decoding information without visual contact between code and decoder. So, in many cases, an extra layer of the polymer could be placed on top to completely cover the implemented patterned code. So, in this context, is relevant to determine the range of distances z where the proposed prototype permits the decoding of the information. Fig. 9 shows how the increase of z led to a diminution in the detection capacity, in terms of lower differences between the voltage amplitudes at the maximums and the minimum observed between magnetic strips. In any case, the code can be read until a limit distance of 0.8 mm between the sensor and top surface, revealing the existence of a broad interval where the code is detectable.

Since the employed MNPs are within the limit of superparamagnetic behavior [26] and taking into account the occurrence of particle size dispersion, some nanoparticles could display a certain magnetization at the remanence after being suitably magnetized. To confirm this, the 3D piece was previously submitted to different external magnetic fields, H_{elec} , along the strips axis (positive sense of the *x*-axis direction, see Fig. 3) to analyze its effect in the detection (see analysis ii) in Section 2.5). The results in Fig. 10 show how the higher the magnetic field applied to the 3D piece, the higher the difference between the two



Fig. 9. Variation of the output signal, *V*_{out}, versus position, *y*, under different distances *z* between sensor and piece surface.



Fig. 10. Variation of the output signal, *V*_{out}, versus position, *y*, under different magnetic fields previously applied to the piece.

maximums and the minimum between the magnetic strips. This behavior can be interpreted in terms of the previous magnetization of the MNPs, and the achievement of higher remanent magnetization of the MNPs assembly as H_{elec} was increased. In fact, as the inset of Fig. 1 shows, the increase in the maximum applied magnetic field in the MNPs hysteresis loops, gave rise to an increment of the magnetization at the remanent state. It is important to note that the chosen magnetic fields in these minor hysteresis loops are the same as those applied by the electromagnet. In consequence, a gradually increasing MNPs stray field acted on the sensing wire, enabling an improvement in the decoding process. So, from a practical point of view, a permanent magnet can be incorporated into the proposed potential prototype to improve and ease the decoding process.

The noticed detection improvement in Fig. 10 would confirm the GMI effect as the main origin in the detection of the MNPs through their stray magnetic field. Nevertheless, to discard other contributions in the detected changes of the wire impedance (i.e. changes in the mutual inductance), the evolution of V_{out} was characterized under the simultaneous effect of H (Helmholtz coils) and the embedded MNPs in the 3D piece after being magnetized in the electromagnet at $H_{elec} = 100 \frac{kA}{m}$. For that, the sensing wire was aligned over the center of strip 1 (*x*-direction) and the whole system was placed in the center of the Helmholtz coils. Keeping the wire fixed, two different positions of the 3D piece were



Fig. 11. Evolution of V_{out} versus external magnetic field under the absence of MNPs (calibrating) and the effect of embedded MNPs in strip 1 at initial (0°) and rotated position (180°). Inset. Zoom of the close to H = 0 region.

analyzed, i.e. the initial (0°) and after its 180° rotation. As is shown in the inset of Fig. 11, an almost symmetrical shift of V_{out} , close to 15 A/m, with respect to the initial calibrating curve (without the effect of the MNPs) was detected. More interesting is the change in the displacement (positive or negative) depending on the orientation of the MNPs orientation respecting the sensor axis. This displacement should be interpreted as an effective magnetic field acting on the sensor, superimposed on the externally applied, *H* by the Helmholtz coils. The observed dependence of V_{out} on the previous magnetizing field and especially on the *direction* of the magnetic strips joined to the contactless detection procedure, justify that the detection is the result of the effect of the MNPs magnetic stray field on the sensing wire impedance.

A similar behavior was observed in a previous work [25] where a detection platform for the contactless and reusable detection of magnetic nanoparticles was developed. In this study, the larger size of the employed MNPs (mean diameter of 140 nm) led to a larger MNPs remanence after the application of the magnetizing field, H_{elec} and so, more favorable detection conditions. In this work, the lower mean size of the targeted MNPs proposed a more exigent scenario, that, enabled to test the proposed system under tougher detection conditions. For that reason, it was required an initial treatment of the sensor nucleus to enhance its response at low external magnetic fields, H, as those generated by the MPNs.

Finally, this work intends to demonstrate whether a low-cost and versatile codification-decoding based on the coupling of the 3D printing technique and the GMI effect can be developed. So, from this initial approach, a subsequent optimization process must be tackled to enable the whole device size reduction (magnetic and polymeric strips and sensor). This can be achieved from a double perspective, namely, i) improvement of the magnetic signal of the embedded MNPs by analyzing different compositions, sizes, etc. searching for larger remanences... and ii) enhancement of the GMI sensing element performance where, as previously mentioned, the use of non-linear effects and/or GMI phase-based devices in differential sensing arrangements would permit to gain sensitivity of detection even at the low required sensing element lengths.

4. Conclusions

A magnetic binary encoding system using 3D printing technology is developed. For this purpose, commercial magnetite nanoparticles were embedded in a polymeric matrix of Polylactic Acid (PLA) and Poly- ϵ -caprolactone (PCL). Then, a printable 3D filament was fabricated and successfully employed for printing a 3D barcoded piece with magnetically encoded information.

A GMI-based prototype decoder has been proposed, based on an amorphous soft magnetic wire of nominal composition ($Co_{0.94}$ Fe_{0.06})_{72.5} Si_{12.5} B₁₅. Besides, for the sensor nucleus excitement and signal conditioning, a homemade low-cost electronic was designed. The complete decoding system has demonstrated the capacity of detecting the low presence of the embedded MNPs (5% wt) even without any previous magnetization and in a moderately long range of distances between the sensor and the top piece surface. Additionally, the detection capacity can be improved with a previous magnetization of the encoded piece. Finally, the GMI effect involvement in the decoding process is confirmed.

A low-cost complete magnetic coded-decoded system is proposed, demonstrating the feasibility of the employed coupled techniques for addressing both the manufacturing and decoding stages. This first approximation paves the way for further studies to optimize the 3D manufactured pieces, the GMI sensing wire and homemade electronics performance to tackle the reduction of the elements and the time of detection.

CrediT authorship contribution statement

Juan Jesús Beato López: Conceptualization, Methodology, Measurements, Validation, Formal analysis, Investigation, Data curation, Writing – original draft, Writing – review & editing. José María Algueta Miguel: Development of Electronic sensor interface, Writing – review & editing. Itziar Galarreta Rodriguez: 3D filament design, fabrication and characterization, Writing – review & editing. Eneko Garaio: 3D filament design and characterization, Software. Alberto López: 3D filament design and characterization. Cristina Gómez Polo: Data curation, Formal analysis, Writing review and editing. Maite Aresti: 3D piece printing, Project Administration, Writing – review & editing. Eneko Soria: 3D piece printing, Data curation. Ignacio Pérez de Landazábal: Conceptualization, Methodology, Validation, Resources, Funding acquisition, Project Administration, Writing – review & editing.

Declaration of Competing Interest

The authors declare that they have no known competing financial interests or personal relationships that could have appeared to influence the work reported in this paper.

Data Availability

No data was used for the research described in the article.

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